Advanced Magnetics Practical Circuit Models

Webinar May 17, 2023 10 am Pacific Coast Time

Dr. Ray Ridley

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Magnetics Circuit Modeling – Some of the Players



Arthur Nace –aerospace engineer and programmer who automated LTspice models for us and made this work possible. Our longest user of RidleyWorks. Dr. Vatché Vorpérian – I've had lots of discussions with him on core loss and proximity loss equivalent circuits. (He produced fractal core loss models back in the 1980s.)





Dr. Qichen Yang who solved the math for 5th and 7th-order circuit models for us.

Dr. Shilpa Marti who spent many weeks collecting core loss information (so that you don't have to.)





Magnetics – Things to Feel Good About



Richard Feynman didn't know how magnets really work either. A great video to watch.

https://www.youtube.com/watch?v=luHDCsYtkTc&ab_channel=AceMon

19h •••



Matthew Kallicharran (Autodidact/Polymath) • 1st Physicist/Engineer - Integrated Magnetics, Embedded Power Converter De...

"We don't actually understand the universe itself. The best we can do is create a model of it and try to understand that..."

There are ZERO equations in this webinar.

Every equation you might need is in the RidleyWorks software.

At the end of this webinar, we will tell you how to get your free copy



Semiconductor Models



Die Size: 0.85 mm x 1.2 mm





.param scale=40.8/25

.param aWg=scale*25 A1=scale*1.593 k2=2.03 k3=0.177 rpara=0.1412/scale

- + alTc=0.0036 arTc=-0.0036 x0_0=1.02 x0_1=0.1789 x0_1_TC=0.004
- + dgs1=scale*4.3e-7 dgs2=scale*2.6e-13 dgs3=.8 dgs4=.23
- + ags1=(scale*15.6E-12 + 13.1E-12) ags2=scale*8.875E-12 ags3=1.56 ags4=0.26
- + ags5=scale*-2.01E-13 ags6=-7.99 ags7=2.46
- + agd1=scale*0.167E-12 agd2=scale*3.523E-12 agd3=-0.889 agd4=1.044
- + agd5=scale*1.14E-12 agd6=-5.658 agd7=4.445
- + asd1=(scale*2.72E-12 + 1.9E-12) asd2=scale*9.00E-12 asd3=-6.1585 asd4=3.1215
- + asd5=scale*6.37E-12 asd6=-42.978 asd7=28.23

rd drainin drain {(0.75*rpara*(1-arTc*(Temp-25)))} rs sourcein source {(0.25*rpara*(1-arTc*(Temp-25)))} rg gatein gate {(.4)}

Rcsdconv drain source {100000Meg/aWg} Rcgsconv gate source {100000Meg/aWg} Rcgdconv gate drain {100000Meg/aWg}

bswitch drain source I=if(v(drain,source)>0,

- + (A1*(1-alTc*(Temp-25))*log(1.0+exp((v(gate,source)-k2)/k3))*
- + v(drain,source)/(1 + max(x0_0+x0_1*(1-x0_1_TC*(Temp-
- 25))*v(gate,source),0.2)*v(drain,source))),
- + (-A1*(1-alTc*(Temp-25))*log(1.0+exp((v(gate,drain)-k2)/k3))*
- + v(source,drain)/(1 + max(x0_0+x0_1*(1-x0_1_TC*(Temp-

25))*v(gate,drain),0.2)*v(source,drain))))

bgsdiode gate source I=if(v(gate,source)>10,

- + (0.5*aWg/1077*(dgs1*(exp((10.0)/dgs3)-1)+dgs2*(exp((10.0)/dgs4)-1))),
- + (0.5*aWg/1077*(dgs1*(exp((v(gate,source))/dgs3)-1)+dgs2*(exp((v(gate,source))/dgs4)-1)))
-)

bgddiode gate drain I=if(v(gate,drain)>10,

- + (0.5*aWg/1077*(dgs1*(exp((10.0)/dgs3)-1)+dgs2*(exp((10.0)/dgs4)-1))),
- + (0.5*aWg/1077*(dgs1*(exp((v(gate,drain))/dgs3)-1)+dgs2*(exp((v(gate,drain))/dgs4)-1))))

C CS	acts sources (cast) TC-0.0
	gate source $(ags1) = 0,0$ gate source $O = (0.5 \times ags2 \times ags4 \times log(1 + oxp(1)/gate source) + ags2)/ags4))+$
C_C031	gate source Q = (0.5 ags2 ags4 log(1+exp((v(gate,source)-ags5)/ags4))+
Ŧ	agso agso log(1+exp((v(source,urain)-agso)/agso))
C GD	gate drain {agd1} TC=0.0
C CGD1	gate drain Q=(0.5*ags2*ags4*log(1+exp((v(gate.drain)-ags3)/ags4))+
+	agd2*agd4*log(1+exp((v(gate.drain)-agd3)/agd4))+agd5*agd7*log(1+exp((v(gate.drain)-
agd6)/agd7)))	
	course drain (acd1) TC=0.0
C_3D	
C_CSD1	source drain Q=(asd2*asd4*log(1+exp((v(source,drain)-asd3)/asd4))+

source drain Q=(asd2*asd4*log(1+exp((v(source,drain)-asd3)/asd4))+ asd5*asd7*log(1+exp((v(source,drain)-asd6)/asd7)))

Magnetics Models





Magnetics Models RL1 L1 **MODEL 1** 000 15.0µ 2.21m M1 Rgate B2 Vcc 10 C3 C2 R3 80 V - 130 V Vin A V=V(drv) 14 130p 12.10K 82.0p RL1 L1 24 V R2 C1 20.00K 2.40n ea out^{vcc} out **15.0μ** -Drv 80.0 RidleyWorks Out PWM SINGLE 2.21m R1 274K Cesr 10m A ea ramp 2 I(RI1) ່ໄວເ .4 14.0A-12.6A-11.2A-9.8A-+8.4A-7.0A-5.6A-4.2A-2.8A-1.4A-0.0A 3.92ms 3.98ms 3.90ms 3.94ms 3.96ms 4.00ms





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Adding Dowell's Equation Results [1]



Linear Plot





Adding Dowell's Equation Results [1]

RL1 L1 Rac MODEL 2 2.21m 15.0µ Helical Foil 3.8mm 1 Layer DOWELL'S Rac CALCULATED PROXIMITY VALUE

Logarithmic Plot





Simulation Results with Dowell's









Adding Core Loss [2]

[2] Magnetics Core Loss Webinar



SER 2918H-153 Losses -130 V Loss (W) 4 3 Coilcraft DC-DC Optimizer 2 Dowells AC + Core 1 **Dowells AC** 0 DCR Only 140 180 220 260 300 **Frequency kHz**





Measuring Rac with the AP310 Analyzer







Thevenin connection for low impedances below 1 mOhm

"Correcting" Measured Rac





Comparison of Different Winding Resistances (Linear Scale)





Comparison of Different Winding Resistances (Logarithmic Scale)



Interesting +1 slope. Where does this come from?

Let's see if it works first....





DCR Loss

♥ Waveform: V(N008,N010)*I(RL1) ×

Interval Start:	1.36ms
Interval End:	1.84ms
Average:	231.36mW
Integral:	111.05µJ

Core Loss

Interval Start:	1.36ms
Interval End:	1.84ms
Average:	437.46mW
Integral:	209.98µJ

ACR Loss

♥ Waveform: V(N011)*lx(Rac_Ind:in) ×				
Interval Start:	1.36ms			
Interval End:	1.84ms			
Average:	1.1507W			
Integral:	552.34µJ			



Final Simulation Results



I'm happy! Are you?

Extra Rac is Due to the Gap







Additional Learning Material

[1] Magnetics Winding Loss

Learn how the winding loss circuit used in today's presentation are generated automatically by RidleyWorks from Dowell's equation solutions.

[2] Magnetics Core Loss

Learn how the core loss circuit used in today's presentation are generated automatically by RidleyWorks

[3] Coilcraft DC-DC Optimizer

Nice tool to get the inductor data from. Maybe they'll have some LTspice models soon.

[4] Francisco de Leon, Adam Semlyen, "TIME DOMAIN MODELING OF EDDY CURRENT EFFECTS FOR TRANSFORMER TRANSIENTS" IEEE Transactions on Power Delivery, Vol. 8, No. 1, January 1993.

Thanks to Jimmy Daniel for verifying some of the simulation results for us



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essential step of product

development. Learn from

our 40+ years experience

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loop gains, impedances

and more.

measurement is an

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I was going thru a design app note from Infineon for a PSFB converter. I was confused by the turn-off loss calculation.

At beginning of mode 3, switch D was conducting. Then it is turned off and the inductor current will discharge Switch C cap and charge Switch D cap. I thought that if Switch D is turned off fast, the charging capacitive current will not dissipate any heat loss on Switch D. It will just be part of the total system lossed.... See more



Since it's a ZVS converter, turn-on loss and output capacitance Co

Turn-off time and loss are:

$$t_{off} = Q_{gd} \cdot \frac{R_g}{V_{pl}} + Q_{gs} \cdot \frac{V_{pl} - V_{th}}{V_{pl}} \cdot \frac{2 \cdot R_g}{V_{pl} + V_{th}}$$
$$= 22 \cdot 10^{-9} \cdot \frac{3}{6.4} + 7 \cdot 10^{-9} \cdot \frac{6.4 - 4}{6.4} \cdot \frac{2 \cdot 3}{6.4 + 4} = 11.83 \cdot 10^{-9} \cdot \frac{10^{-9}}{10^{-9}} \cdot \frac{10^{-9}}{10^{-$$

$$P_{\text{soff}} = 0.5 \cdot I_{I,nk} \cdot \frac{N_s}{\cdots} \cdot V_{in} \cdot t_{off} \cdot f$$

What is the "go to" book that everyone has (and recommends) for transformer winding & design? I am interested in line frequency transformers but if the book also covers SMPS (kHz) types that would be good.



